

## **A NOVEL SET OF THREE LEVEL CONCURRENT WORD LINE BIAS CONDITIONS FOR A NOR TYPE FLASH MEMORY ARRAY**

This application claims priority to Provisional Patent Application serial number 60/271644, filed on Feb. 27, 2001, herein incorporated by reference.

### **BACKGROUND OF THE INVENTION**

#### **1. Field of the Invention**

The present invention relates to semiconductor memories and in particular a three level concurrent word line bias condition for NOR type flash memory arrays.

#### **2. Description of the Related Art**

In today's flash EEPROM technology, a plurality of one-transistor EEPROM cells has been configured into either NAND-type or NOR-type memory arrays. For the NAND type cell array, the sources and drains of the flash cells are connected in series to save die size for the reason of cost reduction. In contrast, for NOR-type cell array, the drains and sources of the cells are connected in parallel to bit lines and source lines, respectively, to achieve high read speed at sacrifice of the increase in die size. It is well known that the NAND-type cell array suffers no over-erase problem due to its unique array structure allowing no leakage path during read. For a one transistor (non-split gate) NOR-type cell array, the over erase problem may or may not occur, and the over erase problem is subject to the choice of erase and program methods. Conventionally, a program operation is performed on the basis of bit-by-bit method but erase is performed collectively on all cells in a block. In both the NOR-type or NAND-type flash memory, the entire flash chip is divided into several blocks, and typically, the size of each flash block ranges from 64Kbits to 512Kbits. An erase operation is performed prior to

program operation, and in a NAND-type flash memory, the erase is performed on a block (sector) basis and program is performed on a page basis.

A page is usually defined as a word line and a block is defined as many word lines which share common bit lines within the same divided block. Although several methods of erase and program operations have been proposed, in the current NAND type flash memory, the definition of erase and program operations is unified. The erase operation is to decrease the  $V_t$  (threshold voltage) of the cells that are physically connected to the same erased word line or the word lines in the same block. In contrast, the program operation is to increase the  $V_t$  of cells of selected erased word line or word lines in the selected block. The non-selected cells in the non-selected word lines in the selected block or the non-selected blocks remain undisturbed.

The following U.S. patents of prior art are directed toward the detailed description of NAND type flash EPROM's.

A) US 6,038,170 (Shiba) is directed toward a nonvolatile memory of a hierarchical bit line structure having hierarchical bit lines which includes a plurality of sub-bit lines.

B) US 5,464,998 (Hayakawa et al.) is directed toward a non-volatile semiconductor memory device including NAND type memory cells arranged in a matrix pattern over a semiconductor substrate.

Up to the present, the definition of erase and program operations for a one-transistor cell, NOR-type flash memory is inconsistent. Erase could be defined to

increase cell's  $V_t$  and program to decrease cell's  $V_t$ , or vice versa depending on the preferred flash technology and its design techniques. The following is a summary of erase and program operations for state of the art one-transistor (non-split-gate) NOR-type flash EEPROM technologies.

1) FN (Fowler-Nordheim) Block erase, CHE (channel hot electron) program, one-transistor cell, NOR-type flash, EEPROM technology. The typical example is an ETOX flash cell. In this prior art, programming is performed on bit-by-bit basis to increase the  $V_t$  of the cells by using the CHE method while erase is performed on block basis to decrease the  $V_t$  of the cells by using FN-tunneling method. The CHE program consumes more than 300uA per bit, therefore only a few bits can be programmed at a time by an on-chip charge pump having an economic semiconductor area. Unlike CHE, FN-tunneling erase requires only 10nA per flash cell so that a big block size of 512Kb can be erased simultaneously. For a  $V_{dd}$  voltage of 3V or lower, about 4 bits of ETOX cells are programmed in state-of-the art design. In a CHE operation, hot electrons are injected into cell's floating gate with an increase in  $V_t$ . In contrast, in the FN tunneling operation, the electrons are extracted out of the floating gate with a decrease in  $V_t$ . The erase operation is called an edge erase operation which is done at edge of the thin tunnel oxide between the floating gate and the source junction. In the ETOX flash cell, the source junction of N+ is used for an erase operation only which is made to be much deeper than the drain node. The source junction of N+ is surrounded with lightly doped N-implant to reduce the peak electrical field generated during erase operation at the tunneling edge. The drain junction is formed with a shallow N+, with a P+ implanted underneath to enhance the electrical field for CHE program. The ETOX cell is made

non-symmetrical with respect to source and drain junctions of the cell in terms of cell structure and operating conditions; therefore, it is very difficult to shrink the cell using technology below 0.18 $\mu$ m for Ultra-high integrated memory.

The key operating conditions for the ETOX technology with a cell made on a P-substrate are as follows:

	<u>Source</u>	<u>Gate</u>	<u>Drain</u>	<u>Bulk</u>
a) Erase (edge)	+5V	-10V	Floating	OV
b) Program (channel)	OV	+10V	+5V	OV
c) Read	OV	Vdd + $\delta$ V	1V	OV
	<u>Erase</u>	<u>Program</u>		
d) Current per cell	10nA	>300uA		

The drawbacks of the ETOX flash cell are: a) a low cell scalability resulting from an asymmetrical cell structure with a deep source junction; b) a high program current caused by the CHE program scheme; c) a high erase current resulting from using an edge-FN scheme with large substrate leakage current; d) severe over erase potential caused by decreasing the  $V_t$  of cells during erase operation; e) a channel punch through problem in short channel lengths due to the edge erase.

The following U.S. patents of prior art are directed toward the detailed description of ETOX flash cell operations:

- A) US 5,712,815 (Colin et al.) is directed toward an improved programming structure for performing a program operation in an array of multiple bits-per-cell flash EEPROM memory cells is provided.
- B) US 5,790,456 (Haddad) is directed toward an improved method for performing channel hot-carrier programming in an array of multiple bits-per-cell Flash EEPROM memory cells in a NOR memory architecture so as to eliminate program disturb during a programming operation.
- C) US 6,011,715 (Pasotti et al.) is directed toward a programming method for a nonvolatile memory which includes several steps that are repeated until a final threshold value is obtained.
- D) US 5,825,689 (Wakita) is directed toward a nonvolatile semiconductor memory device including a memory cell array in which the threshold voltage of a transistor constituting the memory cell is at ground potential or less, and the source voltage condition is changed by a source potential setting circuit in accordance with a detection result from a data detecting circuit.

II) AND one-transistor cell, NOR-type flash EEPROM technology. Unlike ETOX technology, in the AND one transistor prior art the program is performed on bit-by-bit basis to decrease the  $V_t$  of cells while erase is performed on block basis to increase the  $V_t$  of cells. Both erase and program operations use the FN-tunneling method which

consumes only about 10nA per bit; therefore, a large number of flash cells within a large block can be erased simultaneously by an on-chip charge pump which utilizes a small area on the chip. For a single low power supply,  $V_{dd}$ , is at 3V or below, and as many as 16Kb of cells of the AND technology in a block can be erased. In the AND prior art, the erase operation is carried out by FN block channel erase, and the program operation is carried out by page FN edge program. The edge program is at the drain edge formed by a buried N+ bit line. The electrons are injected into cell's floating gate by block channel erase operation with an increase in the  $V_t$  of the erased cells. In contrast, electrons are extracted out of a floating gate by a page edge program operation where the  $V_t$  of the programmed cells decreases. In this AND flash cell, the N+ drain junction is used for program operation only and is made to be much deeper than the source node. The N+ drain junction is surrounded with a lightly-doped N-implant to reduce the peak electrical field that is generated during the drain-edge-program operation. The source junction is formed with a shallow N+ having a P+ implant underneath to prevent voltage punch-through in a short channel region during an edge program operation. The AND cell like the ETOX cell is made non-symmetrical with respect to the source and drain junctions in terms of cell structure and operating conditions. Therefore, it is very difficult to shrink the AND cell below 0.18 $\mu$ m technology for an ultra-high integrated memory.

The key operating conditions for this technology with cell made on P-substrate are summarized as follows.

Source      Gate      Drain      Bulk

a) Erase (channel)	+OV	+15V	OV	OV
b) Program (drain edge)	+5V	-10V	Floating	OV
c) Read	OV	Vdd	1V	OV

	<u>Erase</u>	<u>Program</u>
d) Current per cell	10pA	10nA

The drawbacks of the AND flash cell are: a) low cell scalability caused by asymmetrical cell structure with a deeper drain than source junction; b) high program current resulting from the edge-FN program scheme with large substrate leakage current; c) severe channel punch-through problem in shorter channel length resulting from the edge program.

The detailed description of AND flash cell operations can be referred to the following U.S. patents of prior art:

A) US 6,072,722 (Hirano) is directed toward programming and erasing a non-volatile semiconductor storage device.

B) US 6,101,123 (Kato et al.) is directed toward programming and erasing verification of a non-volatile semiconductor memory.

C) US 6,009,016 (Ishii et al.) is directed toward a nonvolatile semiconductor memory which recovers variation in the threshold of a memory cell due to disturbance related to a word line.

D) US 5,982,668 (Ishii et al.) is directed toward a nonvolatile semiconductor memory which recovers variation in the threshold of a memory cell due to disturbance related to a word line. The nonvolatile memory continuously performs many writing operations without carrying out single-sector erasing after each writing operation.

E) US 5,959,882 (Yoshida et al.) is directed toward a nonvolatile semiconductor memory device with a plurality of threshold voltages set so as to store multi-valued information in one memory cell entitled.

F) US 5,892,713 (Jyouno et al.) is directed toward a configuration that provides a nonvolatile semiconductor memory device which allows high-speed block reading.

G) US 5,757,699 (Takeshima et al.) is directed toward the programming of a selected memory cell which is repeated until the programmed threshold voltage is not greater than a predetermined threshold voltage.

III) FN-erase, FN-program, Metal-bit line, One-transistor, NOR-type Flash EEPROM. Like AND flash technology, in this prior art, the program operation is performed on a bit-by-bit basis to decrease the  $V_t$  of cells while erase is performed on a



block basis to increase the  $V_t$  of cells. Both erase and program operations use the FN-tunneling method, which consumes only about 10nA per bit without taking the greater substrate current into account. Therefore a large number of flash cells within a big block can be erased at one time by an on-chip charge pump having economic area. For a single low power supply,  $V_{dd}$ , of 3V or below, a larger number of flash cells in a block can be programmed and erased simultaneously. In the prior art, the erase operation is carried out by FN channel-erase, and the program operation is carried out by FN edge-program. The edge-program is at the drain edge but the cell structure is formed by a non-buried N+ bit line and a source line. The bit line is a vertical metal line which connects all drains of the cells in the same block for high read speed. The source lines are tied together by an N+ active line, which runs in parallel to the word lines. Each source line is shared by one pair of word lines as in the ETOX flash cell array. As disclosed in the prior art, the electrons are removed from the floating gate of the cells by drain edge FN programming in which the  $V_t$  is decrease. Conversely, the electrons are injected into the floating gate by channel erasing where  $V_t$  is increased. The N+ drain junction is used for the FN program operation and is made to be much deeper than source node, and is surrounded with a lightly doped N-implant to reduce the peak electrical field generated during drain edge program operation. The source junction is formed with shallow N+ with a P+ implant underneath the source to prevent voltage punch-through in a short channel region during edge-program operation. The flash cell of prior art is made asymmetrical with respect to source and drain junctions in terms of cell structure and operating conditions; therefore, it is difficult

to further shrink the memory cell for Ultra-high density memory below 0.18 $\mu$ m technology.

The key operating conditions for the NOR type flash technology with a cell formed on a P-substrate are as follows:

	Source	Gate	Drain	Bulk
a) Erase (channel)	+0V	+15V	0V	0V
b) Program (drain-edge)	Floating	-10V	+5V	0V
c) Read	0V	Vdd	1V	0V
	Erase	Program		
d) Current per cell	10 pA	10nA		

The drawbacks of the NOR type flash cell are: a) Low cell scalability as a result of an asymmetrical cell structure with the drain junction deeper than the source junction; b) high program current caused by the edge-FN program scheme with a large substrate leakage current.; c) severe channel punch-through problem in shorter channel lengths caused by the edge program. The detailed description of the NOR type flash technology can be referred to in U.S. patent 5,708,600 (Hakozaki et al.) which is directed toward a method for writing a multiple value into a nonvolatile memory capable of multiple value data being written into a floating gate type memory cell.

### SUMMARY OF THE INVENTION

An object of the present invention is to provide a three level concurrent word line bias condition and method using CHE program and FN block erase for a semiconductor nonvolatile device and in particular, for an ETOX one transistor cell, and a NOR-type EEPROM memory array formed on P-substrate.

Another object of the present invention is to provide a three level word line bias condition and methods using FN schemes for both program and erase operations for a one transistor cell, NOR type AND EEPROM memory array formed on P-substrate.

Another objective of the present invention is to provide a three level word line bias condition and methods using CHE program and FN block-erase for a semiconductor nonvolatile device, in particular an ETOX one transistor cell, NOR type EEPROM memory array formed on p-well which is within a deep N-well on top of p-substrate.

Still another objective of the present invention is to provide a new operation method that employs the three level word line bias condition to perform the bit-by-bit verify and correction for achieving both tight "0" and "1" for distributions of  $V_t$ .

### BRIEF DESCRIPTION OF THE DRAWINGS

This invention will be described with reference to the accompanying drawings, wherein:

FIG. 1 is a sectional view of an ETOX type memory cell of prior art with a p+ implant and a lightly doped n- implant,

FIG. 2 is a sectional view of an ETOX type memory cell of prior art with a p+ implant,

FIG. 3 is a sectional view of an AND type memory cell of prior art with a p+ implant and a lightly doped n- implant,

FIG. 4 is a sectional view of an NAND type memory cell of prior art with a shallow source and drain,

FIG. 5 is a sectional view of an ETOX type memory cell of prior art with a p+ implant on a p-well within a deep n-well on p-substrate,

FIG. 6a through 6e show a single cell operating conditions of the present invention for an ETOX NOR flash cell array on a P-substrate,

FIG. 7a through 7d show additional single cell operating conditions of the present invention for an ETOX NOR flash cell array on a P-substrate,

FIG. 8 shows the ETOX NOR flash cell array of prior art,

FIG. 9 illustrates the bias conditions for block erase for the ETOX NOR flash cell array of the present invention,

FIG. 10 illustrates the bias conditions for block erase verify for the ETOX NOR flash cell array of the present invention,

FIG. 11 illustrates the bias conditions for erase inhibit for the ETOX NOR type flash cell array of the present invention,

FIG. 12 illustrates the bias conditions for correction operations for the ETOX NOR type flash cell array of the present invention,

FIG. 13 illustrates the bias conditions for correction verify operation for the ETOX NOR type flash cell array of the present invention,

FIG. 14 illustrates the bias conditions for program operations for the ETOX NOR type flash cell array of the present invention,

FIG. 15 illustrates the bias conditions for program verify operations for the ETOX NOR type flash cell array of the present invention,

FIG. 16a through 16e show a single cell operating conditions for AND arrays on P-substrate for the present invention,

FIG. 17a through 17f show additional single cell operating conditions for AND arrays on P-substrate for the present invention,

FIG. 18 shows an AND flash cell array of prior art,

FIG. 19 illustrates the bias conditions for random page erase operation of the present invention for the AND flash cell array of the present invention,

FIG. 20 illustrates the bias conditions for random page erase verify operation of the present invention for the AND flash cell array of the present invention,

FIG. 21 illustrates the bias conditions for block erase operation of the present invention for the AND flash cell array of the present invention,

FIG. 22 illustrates the bias conditions for block erase verify operation of the present invention for the AND flash cell array of the present invention,

FIG. 23 illustrates the bias conditions for block erase inhibit operation of the present invention for the AND flash cell array of the present invention,

FIG. 24 illustrates the bias conditions for correction operation of the present invention for the AND flash cell array of the present invention,

FIG. 25 illustrates the bias conditions for correction verify operation of the present invention for the AND flash cell array of the present invention,

FIG. 26 illustrates the bias conditions for random page program operation of the present invention for the AND flash cell array of the present invention,

FIG. 27 illustrates the bias conditions for random page program verify operation of the present invention for the AND flash cell array of the present invention,

FIG. 28a through 28e show a single cell operating conditions for ETOX NOR arrays on a P-well for the present invention,

FIG. 29a through 29d show additional single cell operating conditions for ETOX NOR arrays on a P-well for the present invention,

Fig. 30 illustrates an ETOX NOR flash cell array on a P-well of prior art,

FIG. 31 illustrates the bias conditions for block erase operations of the present invention for the ETOX NOR array of prior art on a P-well,

FIG. 32 illustrates the bias conditions for block erase verify operations of the present invention for the ETOX NOR array of prior art on a P-well,

FIG. 33 illustrates the bias conditions for erase inhibit operations of the present invention for the ETOX NOR array of prior art on a P-well,

FIG. 34 illustrates the bias conditions for correction operations of the present invention for the ETOX NOR array of prior art on a P-well,

FIG. 35 illustrates the bias conditions for correction verify operations of the present invention for the ETOX NOR array of prior art on a P-well,

FIG. 36 illustrates the bias conditions for program operations of the present invention for the ETOX NOR array of prior art on a P-well,

FIG. 37 illustrates the bias conditions for program verify operations of the present invention for the ETOX NOR array of prior art on a P-well,

FIG. 38a shows the  $V_t$  distribution obtained after application of the block erase and page program sequence of operations in an ETOX NOR array of prior art,

FIG. 38b shows the  $V_t$  distribution obtained after application of the block erase and page program sequence of operations of the present invention to the cells in an ETOX NOR array,

FIG. 39 is a flow diagram of the present invention for block erase operations in an ETOX NOR array,

FIG. 40 is a flow diagram of the present invention for correction operations of an ETOX NOR array,

Fig. 41a shows a  $V_t$  distribution obtained after application of block erase and page program sequence of operations for a first AND cell array of prior art,

Fig. 41b shows a  $V_t$  distribution obtained after application block erase and page program sequence of operations for a second AND cell array of prior art,

Fig. 41c shows a  $V_t$  distribution obtained after application of block erase and page program sequence of operations of the present invention to an AND cell array,

Fig. 42a is a flow diagram block erase and page program sequence of operations of prior art applied to a first AND array of prior art,

Fig. 42b is a flow diagram block erase and page program sequence of operations of prior art applied to a second AND array of prior art,

FIG. 43 is a flow diagram of the present invention for block erase operations for an AND array of the present invention, and

FIG. 44 is a flow diagram of the present invention for correction operations for an AND array of the present invention.

### DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

The operating principles of a one-transistor flash cell of the present invention are described with reference to FIG. 1-5. The term "write operation" will be frequently used in this description and is defined as an operation, comprising erase and program operations. In one complete write operation, erase is usually performed first on a block basis followed by a program operation on a page basis.

FIG.1 shows the cross sectional view of a device structure of prior art of an ETOX flash memory cell with an n+ source 33, n+ drain 22, control gate 10 and floating gate 11. The tunnel oxide layer 15 is formed between floating gate 11 and P-substrate 16. The arrow 48 shows the flow of electrons from floating gate 11 to source 33 during an edge erase operation. The arrow 49 shows electrons moving from drain 44 to the floating gate 11 during CHE (channel hot electron) program operation.

Continuing to refer to FIG. 1, since the n+ source node 33 experiences much higher electric field during an edge erase operation than the drain 22 during CHE program operation, the source junction is made much deeper than drain. The source 33



is lightly doped by an n- implant 34 to avoid junction breakdown in the erase operation. The p+ implant 44 is used to increase the substrate concentration underneath n+ drain 22 so that the CHE program operation can be achieved. An n- implant 34 is formed underneath n+ source 33 so that breakdown can be avoided during FN edge erase 48. A second prior art of an ETOX cell is shown in FIG. 2. Shown in FIG. 2 is a cross sectional view of ETOX cell which uses a CHE program 49 to increase the threshold voltage,  $V_t$ , and an FN channel erase 38 to decrease  $V_t$ . In FIG.2, the n- implant layer is no longer required since channel erase 38 does not exert the high tunneling electric field to source junction 13.

In FIG. 3 is shown a cross sectional view an AND flash cell of prior art with buried N+ source 13 and drain 12 which uses FN channel erase 18 to increase  $V_t$  and FN edge program 19 to decrease  $V_t$ . Edge program is performed similarly to the ETOX cell in FIG.1. A lightly doped n- layer 20 is required underneath n+ drain 12 to reduce the high tunneling electric field during program operation. The cell of FIG. 3 is an asymmetric cell and has a lower scalability as with the cell in FIG. 2.

The prior art of FIG. 4 shows a cross sectional view of a NAND like flash cell with an n+ source 13 and drain 22. This flash cell uses FN channel-program 28 to increase  $V_t$  and FN channel-erase 27 to decrease  $V_t$ . Unlike the previous cells, neither n- nor p+ are required for drain 22 and source node 13, respectively. The cell of FIG. 4 is a symmetrical cell and has higher scalability than the cells in the cells shown in FIG. 1, 2 and 3.

The prior art of FIG. 5 shows a cross sectional view of an ETOX cell which uses a CHE program 49 to increase  $V_t$  and a FN channel erase 48 to decrease  $V_t$ . The cell is

formed on p-well 40 within an n-well 41 on a P-substrate 16. The n-implant layer is no longer required since channel erasing 48 does not exert the high tunneling electric field on the source junction 13. For lower voltage operation of this cell the voltages applied to the control gate 10 and p-well 40 can be DC-shifted down.

The first embodiment of the present invention will be described with reference to FIG. 6 through FIG. 15. The cell used is an ETOX cell on P-substrate. FIG. 6a shows a flash cell of this invention with nodes of D, G and S. The P-substrate is not shown and is held at ground. FIG. 6b shows a flash cell with bias conditions of this invention illustrating two types of ERASE operations. The first operation has D, G and S nodes coupled with floating, -10V and +5V, respectively, for edge-erase. The second operation has D, G and S nodes coupled with 0V, -15V and 0V, respectively, for a channel erase, where the -15V is an exemplary value. The exact value and time of the control gate voltage is subject to different flash technologies. The gate voltage of -15V, and source and drain voltage of 0V will result in a tunneling electric field in channel region of the cell. The tunneling electric field will transport electrons from the floating gate to the P-substrate in order to decrease the  $V_t$  of the cells (off-state) after a predetermined erase time. The erase operation can be performed on the basis of single-page (word line), block (N word lines), multiple blocks (M blocks) or chip (all blocks), where typically N and M are larger than 2.

FIG. 6c shows a flash cell with bias conditions of the present invention illustrating two types of an erase inhibit operation. The first erase inhibit is with D, G and S nodes set to floating, -10V, +5V respectively, and the second erase inhibit with D, G and S nodes set to 0V, 0V and 0V, respectively. This operation is intended to prevent a build-

up of disturbance to those non-selected erased cells (in either selected or non-selected blocks) and to achieve better endurance (number of program and erase cycles).

FIG. 6d shows a flash cell with bias conditions of the present invention that illustrates the correction operation with D, G and S nodes coupled to +5V,  $V_{\text{correction}}$  and 0V, respectively. Here  $V_{\text{correction}}$  is an exemplary value for better understanding of the present invention. The exact value and time for the control gate voltage in this operation varies with different flash technologies. The correction operation is a soft program CHE operation. The difference between program and correction lies in the control gate voltage. Normal program operation has about +10V applied to control gate and is intended to increase  $V_t$  more than +5V. Correction has lower control gate voltage to avoid over program. It is used to correct the  $V_t$  of cells back to around +1V from either negative or below +0.5V to avoid sub-threshold leakage during subsequent read or program operations. The operation of FIG. 6d is sometimes referred to as recovery. The data becomes "0" after this operation.

FIG. 6e shows a flash ETOX cell with the bias conditions of the present invention illustrating the CHE program operation with D, G and S nodes coupled to +5V,  $V_{\text{pgm}}$  and 0V, respectively. In the CHE program, there is a conduction current flowing from drain to source and causing an electron-hole pair generated at drain node. Electrons are attracted to the floating gate to increase  $V_t$  by the positive high voltage  $V_{\text{pgm}}$ , which increases the  $V_t$  of the cells. The CHE program typically consumes more than 300uA per cell. With weak on chip charge pump circuits operating at  $V_{\text{dd}}$  below 3V, only about 4 bits can be programmed simultaneously. The cell data becomes "1" after this operation is performed.

In FIG. 7a, a flash cell with bias conditions of the present invention illustrates the program verify operation with D, G and S nodes coupled to +1V,  $V_{pgmvfy}$  and 0V, respectively. The  $V_{pgmvfy}$  is an adjustable voltage input to the control gate of the cells to meet different  $V_t$  requirements in the program operation. For example, for storage of more than 2-bits per cell,  $V_{pgmvfy}$  may vary from as low as 1V up to about 5V or more. For a binary program,  $V_{pgmvfy}$  is set to be around +5V. For storage of multiple states such as 1V, 2V, 3V, and 4V,  $V_{pgmvfy}$  is set to 1V, 2V, 3V and 4V accordingly to verify each state

In FIG. 7b, a flash cell with bias conditions of the present invention illustrates the correction verify operation with D, G and S nodes coupled to +1V,  $V_{corvfy}$  and 0V, respectively. The  $V_{corvfy}$  is an adjustable voltage input to the control gate to meet different  $V_t$  requirements in this operation. The  $V_t$  of over erased cells will be recovered back to a  $V_t$  window of between + 0.5V and 1.0V, after this operation is performed.

FIG. 7c shows a flash cell with bias conditions of the present invention illustrating a read operation with D, G and S nodes coupled to +1V,  $V_{read}$  and 0V, respectively. The  $V_{read}$  is an adjustable voltage input for the control gate for the read operation.  $V_{read}$  can be simply set to  $V_{dd}$ ; however, in some designs,  $V_{read}$  is set to a clamped value so that the read voltage applied to the control gate can be independent of  $V_{dd}$  variation. In some designs, the  $V_{read}$  voltage is boosted to be higher than  $V_{dd}$ , resulting in higher read current.

Shown in FIG. 7d is a flash cell with bias conditions of the present invention illustrating an erase verify operation with D, G and S nodes coupled to +1V,  $V_{ersvfy}$  and 0V, respectively. The  $V_{ersvfy}$  is an adjustable voltage input to the control gate for this

operation. In the conventional ETOX cell,  $V_{ersvfy}$  is set to be around +2.5V to reduce the number of over erased cells because the correction cannot be performed in a bit-by-bit mode as in regular program operation. The correction is done in a collective mode. When the number of over-erased cells increases to some level, the correction current will overload the on-chip charge pump and fail to recover the  $V_t$  of the cells. In contrast,  $V_{ersvfy}$  is set to be +1V in the present invention and as a result of the 3-level word line voltage is used to perform bit-by-bit correction. There are many cells in many bit lines, but only one cell per one selected bit line is corrected simultaneously. Therefore, current over load will not occur and the corrected  $V_t$  can be set accurately.

In FIG.8 is shown a conventional ETOX NOR type flash EEPROM memory array 10. This nonvolatile NOR-type memory array includes: a matrix of word lines and bit lines intersecting one another; and an ETOX memory cell being disposed so as to correspond to each intersection of the matrix of the global bit lines  $BL_n - BL_{n+1}$ , local bit line  $B_n$ , source line  $SL_n$  and global word lines  $WL_n$ , the memory cell including a control gate, a drain, a source and a P-substrate as shown in FIG.1. The control gates are coupled to a corresponding one of the row wise word lines  $WL_m(n)$ , the drains are coupled to a corresponding one of the local column wise bit line  $B_n$  and one of the global bit line  $BL_n$  selected by transistor  $T_n$  gated by  $BT1(n)$  and  $BT2(n)$ , and the sources are coupled to a corresponding one of the local row wise source lines  $SL_m$ . The memory cell is capable of performing a FN erase and a CHE program operation based on the three level word line of the present invention. The plurality of control signals of  $WL_m(n)$ ,  $BT1(n)$  and  $BT2(n)$  are generated from an X-decoder (word

line decoder), local bit line decoder, local source line decoder, global bit line decoder and global source line decoder, respectively.

Shown in FIG.9 are two Block Erase Operations for a conventional ETOX NOR type flash EEPROM memory array 10. One is edge erase operation, and the other is a channel erase operation. For channel-erase operation, an erase voltage of -15V is coupled to the corresponding row wise word lines, WL0(0)-WL511(0) for selected Block 0, and ground is coupled to the non-selected word lines in the rest of the blocks. The drains of the cells are coupled to 0V by a corresponding the local column wise first level metal bit lines B0-B3 and the second global metal bit lines BLn-BLn+1 via transistors of T0 and T1 which are gated by applying Vdd to BT1 (n) and BT2(n) signals. The sources are coupled to 0V by a corresponding plurality of row wise source lines SL0(0)-SLm(0). The channel erase conditions are applied so that the memory cells of Ma -Ml in the word lines in Block 0 are capable of performing a FN erase operation. The Vt of the cells are decreased after the erase operation is performed. The flash cells of Mm-Mx in non-selected Block n are kept undisturbed. The erase operation is designed to be an iterative process. Each erase pulse width can be set to around 1 ms. Given a shorter erase pulse, a larger number of erase pulses are required.

In FIG. 10 is illustrated a Block Erase Verify operation with preferred voltages for WLn(n), global BLn, global SLm, BT1(n), BT2(n). Assuming WL0(0) is firstly selected for Block erase verify, then Versvfy is coupled to WL0(0). The rest of the word lines of WL1(0) to WL511(0) are coupled to -4V to shut off any potential leakage caused by over-erased cells that might exist in Block 0. The word line voltage of -4v is not a fixed number but is set to be able to shut off any leakage current resulted from cells with a Vt

less than -4V. All source voltages  $SL_m(n)$  are coupled to ground. Bit line  $BL_n$  is coupled to 0V and  $BL_{n+1}$  is coupled to 1V for the verify operation. Bit line  $BL_{n+1}$  is then selectively connected to a corresponding sense amplifier for verification. In the conventional ETOX array a total of eight sense amplifiers are needed for byte read and 16 sense amplifiers for word read. The verification for the rest of the cells on the same word lines will be controlled by connecting the sense amplifiers to the next group of 8 bit lines. The process will be continued in the same page until all cells in the page are verified. Then the verify process is moved to the next page of  $WL_1(0)$  in Block 0. After five hundred and twelve word lines are verified, the block erase verify is terminated. With the successful verification of block erase all cells in the selected block (Block 0) become a logical "0". The data of the cells in the non-selected blocks remain the same without changes. Three word line voltages are used concurrently, which include  $V_{ersvfy}$  for the selected word line, -4V for non-selected word lines in the selected block and 0V for non-selected word lines in non-selected blocks. Although the cells in Block 0 are at a logical "0", there could be over erase cells. The definition of an over erased cell in the present invention is a cell  $V_t$  ranging from a negative  $V_t$  to a positive  $V_t$  but below 0.5V. The over-erase cells will induce leakage and result in false readings so that a  $V_t$  correction operation is required.

Fig. 11 shows an erase inhibit operation with preferred voltages for  $WL_m(n)$ , global  $BL_n$ , global  $SL_m$ ,  $BT_1(n)$ , and  $BT_2(n)$ . This operation is performed on sub-block basis and is intended to set those sub-blocks that have been successfully erased into a de-selected mode to prevent further erase. For example,  $WL_0(0)$  and  $WL_1(0)$  are verified to have a successful erase and are set to be in Erase Inhibit mode to avoid the

further erase pluses. The way to set erase inhibit is to set the word line voltage from  $-15\text{V}(\text{erase})$  to  $0\text{V}$  (inhibit) for channel erase or source voltage from  $+5\text{V}$  (erase) to  $0\text{V}$  (inhibit) for an edge erase operation. In the erase inhibit mode the tunneling electric field is reduced so that no tunneling effect will take place. This operation does not require three concurrent word line voltages. In FIG. 12 a correction operation is illustrated with preferred voltages for  $\text{WLm}(\text{n})$ , global  $\text{BLn}$ , global  $\text{SLm}$ ,  $\text{BT1}(\text{n})$  and  $\text{BT2}(\text{n})$ . The correction operation is performed on a bit by bit basis and is intended to correct those over erased cells to a  $V_t$  voltage that is positive but below  $+0.5\text{V}$ . The process is repeated to correct all cells in one selected word line  $\text{WL0}(\text{0})$  and then moved to correct the cells in next word lines of Block 0. A  $V_{\text{correction}}$  voltage is coupled to the first selected word line  $\text{WL0}(\text{0})$  along with a bit line voltage of  $5\text{V}$  in order to perform a CHE soft program. The  $V_{\text{correction}}$  voltage is set to be less than  $+10\text{V}$  while the rest of the word lines of  $\text{WL1}(\text{0})$  to  $\text{WL511}(\text{0})$  are coupled to  $-4\text{V}$  to shut off any potential leakage due to over erased cells that might be existing in Block 0. The  $-4\text{V}$  is an approximate value and is of sufficient magnitude to be able to shut off any leakage current resulting from cells with  $V_t$  less than  $-4\text{V}$ . All source voltages of  $\text{SL0}(\text{0}) - \text{SLm}(\text{n})$  are coupled to ground along with bit line  $\text{BLn}$ . Bit line  $\text{BLn}+1$  is couple to  $5\text{V}$  for the correction operation. The process is continued in the same page until all cells in the page are corrected and then the process moves to correct next page  $\text{WL0}(\text{1})$  in Block 0. After all the word lines in a block are successfully corrected, the correction process is terminated. With the successful correction, all cells in the selected block (Block 0) become "0". The data of cells in the non-selected blocks remain the same without



changes. This correction operation uses three concurrent word line voltages,  $V_{\text{correction}}$ , -4V and 0V.

FIG. 13 illustrates a correction verify operation with preferred voltages for  $WLn(n)$ , global  $BLn$ , global  $SLm$ ,  $BT1(n)$ , and  $BT2(n)$ . This operation is performed on a bit-by-bit basis. In the present invention, this operation is intended to verify that those over erased cells are corrected to  $V_t$  within +0.5V but below +1V. A voltage  $V_{\text{corvfy}}$  is coupled to word line  $WL0(0)$  and the rest of the word lines,  $WL1(0)$  to  $WL511(0)$ , are coupled to -4V to shut off any potential leakage caused by over erased cells that might be existing in Block 0. The voltage of -4v is an approximate value and is set to be able to shut off any leakage current which results from cells with  $V_t$  less than -4V. All source voltages of  $SL0(0) - SLm(n)$  are coupled to ground along with the bit line  $BLn$ . The bit line  $BLn+1$  is coupled to +1V for the verify operation. The process will be continued in the same page until all cells in that page are corrected and verified. Then the operation moves to verify next page of  $WLn(0)$  in Block 0. After all word lines in the block are successfully corrected and verified, the verify process is terminated. Three word line voltages are used, which include  $V_{\text{corvfy}}$  for the selected word line, -4V for non-selected word lines in the selected block and 0V for non-selected word lines in non-selected blocks.

In FIG. 14 is illustrated a CHE program operation with preferred voltages for  $WLn(n)$ , global  $BLn$ , global  $SLm$ ,  $BT1(n)$ , and  $BT2(n)$ . This operation is performed after correction and is on a bit-by-bit basis.. The process is continued until all cells are fully verified. A voltage  $V_{\text{pgm}}$ , which is approximately 10V, is coupled to  $WL0(0)$ , and the remainder of the word lines of  $WL1(0)$  to  $WL511(0)$  are coupled to 0V. Because all over

erased cells have previously been corrected, the  $-4V$  used to shut off any leakage is no longer needed. The CHE program operation performs on a bit-by-bit basis and terminates when all cells in same byte/word are programmed, coupled to a high  $V_t$  ( $>4V$ ). Three concurrent word line voltages are not required for this operation.

In FIG. 15 a CHE program verify operation is illustrated with preferred voltages for  $WLn(n)$ , global  $BLn$ , global  $SLm$ ,  $BT1(n)$ , and  $BT2(n)$ . This operation is performed in a similar manner as correction verify. The only difference is that the verify voltage for the programmed cell,  $V_{pgmvfy}$ , is set to be approximately  $+4V$  for cell  $V_t$  of  $+4V$  after a CHE program operation. Three concurrent word line voltages are not required for this operation.

In summary, there are three preferred operations for an ETOX array that require a three level voltage word line. These operations include block erase verify, correction and correction verify. In the block erase operation, the word line of the selected page within the selected block is couple with  $V_{ersvfy}$ , the word lines in the non-selected pages within the selected block are couple to a voltage approximately  $-4V$ , and the word lines in the non-selected blocks are couple to  $0V$ . In the correction operation the word line in the selected page of the selected block is coupled to  $V_{correction}$ , the non-selected pages in the selected block are coupled to a voltage approximately  $-4V$ , and the word lines in the non-selected blocks are couple to  $0V$ . In the correction verify operation the selected page within the selected block is coupled with  $V_{corvfy}$ , the word lines in the non-selected pages within the selected block are couple to a voltage approximately  $-4V$ , and the word lines in the non-selected blocks are couple to  $0V$ .

The second embodiment of the present invention will be described with reference

to FIG. 16 through FIG. 27 for an AND array on P-substrate by employing the same three voltage word line technique of the present invention for some preferred operations. The erase operations use both edge and channel to perform the FN method. Erase operation is carried out to decrease  $V_t$  of the cells and program increases  $V_t$ . The current using FN program and FN erase only causes 10pA and 10nA, respectively, per cell.

In FIG. 16a is shown a simplified form of a flash cell on P-substrate of the present invention with three nodes of D, G and S. Where D, G and S stand for drain, gate and source, respectively. The potential of P-substrate is held to ground level and is not shown in the figure. In FIG. 16b a flash cell of the present invention shown biased for two types of erase operations with first set of voltages for D, G and S coupled with +5V, -10V and 5V, respectively, for an edge erase, and the second set of voltages for D, G and S coupled with 0V, -15V and 0V, respectively, for channel erase. Where -10V and -15V are exemplary values. The exact value and time of the gate voltage in the erase operation varies with different flash technologies. Gate voltages of -15V/-10V and source and drain voltages of 0V/+5V will result in FN tunneling in the channel region for channel erase and edge region for edge erase. The tunneling electrons will flow from the floating gate to the P-substrate and the source and drain to decrease  $V_t$  of the cell after a predetermined erase time. The erase operation can be performed on the basis of single page (word line), block (more than 2 word lines), multiple blocks (more than two blocks) and chip (all blocks).

In FIG. 16c is shown a flash cell of the present invention biased in two types of erase inhibit operations with D, G and S coupled to a first set of voltages +5V, +5V and

5V, respectively, for an edge erase, and D, G and S coupled to a second set of voltages 0V, 0V and 0V, respectively, for a channel erase. The erase inhibit operation can be performed on the basis of single page (word line), block (more than 2 word lines), multiple blocks (more than two blocks) and chip (all blocks).

FIG.16d shows a simplified form of a flash cell on a P-substrate of the present invention with the three transistor nodes of D, G and S biased with 0V,  $V_{\text{correction}}$  and 0V, respectively, in correction mode. The correction operation is performed on a page basis and is intended to verify the  $V_t$  of all cells in one selected word line after an erase operation.

Referring to FIG.16e, a simplified form of a flash cell is shown on a P-substrate of the present invention with three nodes of the cell transistor D, G and S biased with 0V,  $V_{\text{pgm}}$  and 0V, respectively, in a program operation. This operation is performed on page basis and is intended to simultaneously program the selected cells to high  $V_t$  ( $>4V$ ).

FIG.17a shows a simplified form of a flash cell on a P-substrate of the present invention with three nodes of the cell transistor D, G and S biased with +5V,  $V_{\text{pgm}}/V_{\text{correction}}$  and +5V, respectively, for program and correction inhibit operation. This operation is intended to prevent non-selected cells from programming and correction in the same word line of the selected block.

In FIG.17b a simplified form of a flash cell is shown on a P-substrate of the present invention with three nodes of D, G and S biased with +5V/0V, +2.5V and +5V/0V, respectively, for program operation and 0V, +2.5V, and 0V, respectively, for

correction inhibit. This is intended to prevent non-selected cells on word lines not selected in the selected block from bit line disturb of programming and correction.

In FIG. 17c a flash cell of the present invention is shown biased in a program verify operation with nodes D, G and S coupled with +1V,  $V_{pgmvfy}$  and 0V. This operation can be performed on page basis. FIG. 17d shows a flash cell of the present invention biased in correction verify operation with D, G and S coupled with +1V,  $V_{corvfy}$  and +0V, respectively. This operation can also be performed on page basis. In FIG. 17e a flash cell of the present invention is biased in a read operation with nodes D, G and S coupled with +1V,  $V_{read}$  and 0V, respectively. This operation can be performed on page basis. FIG. 17f shows a flash cell of the present invention biased in an erase verify operation with nodes D, G and S coupled with +1V,  $V_{ersvfy}$  and +0V, respectively. This operation can be performed on page basis.

In FIG. 18 is shown a conventional AND memory cell, NOR-type flash EEPROM memory array 20. This nonvolatile NOR-type memory array includes a matrix of word lines and bit lines intersecting one another. The AND memory cell  $M_a$  to  $M_x$  being disposed so as to correspond to each intersection of the matrix of the global bit lines  $BL_m$  to  $BL_{m+3}$ , local bit line  $B_n$ , source line  $SL_n$  and global word lines  $WL_n$ . The memory cell including a control gate, a drain, a source and a P-substrate is shown in FIG. 1. The control gate is coupled to a corresponding one of the row-wise word lines  $WL_m(n)$ . The drains are coupled to a corresponding one of the local column-wise bit line  $B_n$ , and one of the global bit lines  $BL_n$  is selected by transistor  $T_n$  gated by  $BT1(n)$ . The sources  $S_n$  are coupled to a corresponding one of the local row-wise source line  $SL(n)$  via transistor  $T_n$  gated by  $ST(n)$ , and the memory cell is capable of performing

an FN erase and FN program operations based on the three level word line bias of the present invention. The plurality of control signals of  $WLM(n)$ ,  $BT1(n)$ ,  $ST1(n)$  are generated from X-decoder (word line decoder), local bit line decoder, local source line decoder, global bit line decoder and global source line decoder, respectively.

FIG.19 shows an AND flash array biased in Random page Erase Operation with selected page  $WL0(0)$  coupled with  $-10V$  and  $BLm$  to  $BLm+3$  coupled with  $+5V$  for edge erase. Word line  $WL0(0)$  is coupled with  $-15V$  and  $BLm-BLm+3$  coupled  $0V$  for channel-erase. Word lines  $WL1(0)$  through  $WL31(0)$  are coupled to  $+5V$  for the edge erase operation and coupled to  $0V$  for the channel erase to reduce the erase disturb to those non-selected cells in the selected block, Block 0. The non-selected cells include cells that are in cells  $Ma-Ml$  of Block 0. The nodes of the rest of non-selected cells in the non-selected blocks are all biased with  $0V$  for the drains, gates and sources. The bias condition for the non selected cells prevents erase disturb of cells of  $Mm-Mx$  in non-selected blocks. After the Random page erase operation is performed, the  $V_t$  of cells in  $WL0(0)$  will be decreased and data "1" is stored in the cells.

FIG. 20 shows a random page erase verify operation with preferred voltages for  $WLM(n)$ , global  $BLm$ , global  $SL(n)$ ,  $BT1(n)$  and  $ST1(n)$ . This operation is performed after the completion of a random page erase. In the present invention, this operation is intended to verify the  $V_t$  of those cells erased by verifying that the value of  $V_t$  is below  $+1V$ . This operation can be carried out on page basis using a plurality of sense amplifiers connected to  $BLm$ . Random page is an arbitrary page of any block selected to perform erase operation. A confirmed success by page verify means all cells in the selected page have been erased to be data "0" with  $V_t$  below  $+1V$ . Some fast cells may have

been over erased with  $V_t$  becoming negative which requires a  $V_t$  correction (recovery) in the subsequent operation. In this operation,  $V_{ersvfy}$  of +1V is coupled to the selected word line (page) for data verification. The rest of word lines are coupled to ground level to shut off the bit line leakage. Since only one word line is selected in this operation, no three voltage word line bias is required.

In FIG. 21 a block erase operation is shown with preferred voltages for  $W_{Lm}(n)$ , global  $B_{Lm}$ , global  $S_{L}(n)$ ,  $BT1(n)$  and  $ST1(n)$ . In the present invention, the block erase operation is intended to erase a plurality of cells in a selected block simultaneously. A typical flash block comprises of thirty two word lines and thousands of bit lines. In FIG. 21, cells in Block 0 are selected for erase. Part of the cells in Block 0 include  $M_a, M_b, M_c, M_d$  on word line  $W_{L0}(0)$ ,  $M_e, M_f, M_g, M_h$  on  $W_{L1}(0)$  and  $M_i, M_j, M_k$  and  $M_l$  on  $W_{L31}(0)$ . All word lines from  $W_{L0}(0)$  to  $W_{L31}(0)$  in Block 0 are coupled to -15V with all selected bit lines from  $B_{Lm}$  to  $B_{Lm+3}$  coupled to 0V for channel erase. All word lines of  $W_{L0}(0)$ - $W_{L31}(0)$  in Block 0 are coupled to -10V with all selected bit lines of  $B_{Lm}$  to  $B_{Lm+3}$  coupled with +5V for edge erase. In both cases, source line  $S_{L}(0)$  is held at ground level during the erase operation. In edge erase,  $BT1$  is coupled to +10V to transfer +5V from global bit lines of  $B_{Lm}$  through  $B_{Lm+3}$  to local bit lines of  $B_0$  through  $B_3$ . The source control line  $ST1$  is coupled to 0V to shut off the local source lines of  $S_0$  through  $S_3$  to the common source line of  $S_{L}(0)$  in Block 0. In channel erase,  $BT1$  is coupled to  $V_{dd}$  to transfer 0V from global bit lines of  $B_{Lm}$  through  $B_{Lm+3}$  to local bit lines of  $B_0$  through  $B_3$ . The source control line  $ST1$  is coupled to  $V_{dd}$  to connect the local source lines of  $S_0$  through  $S_3$  to the common source line of  $S_{L}(0)$  in Block 0. Some fast cells may have been over-erased with negative  $V_t$  that require a  $V_t$  correction

(recovery) in the subsequent operation. In this operation, three voltage are not required for the word lines. For the remaining word lines in non-selected blocks are all coupled to ground to avoid any erase disturbance.

FIG. 22 shows a block erase verify operation with preferred voltages for  $W L_m(n)$ , global  $B L_m$ , global  $S L(n)$ ,  $B T_1(n)$  and  $S T_1(n)$ . In the present invention, the block erase verify operation is intended to simultaneously verify a plurality of cells in a selected block on page basis. Each of the thirty two word lines in Block 0 is sequentially selected for data verification. The cells in  $W L_0(0)$  are selected at the same time for page verify. In this operation, word line  $W L_0(0)$  is coupled to Versvfy voltage at approximately around +1V. The rest of word lines in Block 0,  $W L_1(0)$  through  $W L_{31}(0)$ , are coupled with -4V to shut off leakage and avoid false verification. The -4V is used with the assumption that the  $V_t$  of all cells is not more negative than -4V after block erase operation. The non-selected cells in non-selected word lines in non-selected blocks, Block 1 through Block n, are grounded. In this manner, three voltages are used for word lines to achieve a bit-by-bit verify in the present invention. The three word line voltages include Versvfy for the selected word line, -4V for non-selected word lines in the selected block and 0V for non-selected word lines in non-selected blocks. When all cells in  $W L_0(0)$  have been verified, the operation is continued to verify the next word line  $W L_1(0)$ .  $W L_0(0)$  is switched to -4V and  $W L_1(0)$  is coupled to Versvfy. Then, the same steps used to verify  $W L_0(0)$  is repeated on  $W L_1(0)$ . When  $W L_{31}(0)$  has been verified, the process will terminate. After the completion of this operation, all cells in Block 0 are "0" with  $V_t$  less than +1V.



Referring to FIG. 23, a block erase inhibit operation is shown with preferred voltages for  $W L_m(n)$ , global  $B L_m$ , global  $S L(n)$ ,  $B T_1(n)$  and  $S T_1(n)$ . The bias conditions for the block erase inhibit operation is provided to further reduce the erase disturbance to those pages which have passed the erase and erase verify during the iterative erase operation. For example, except for word line  $W L_{31}(0)$ , word lines  $W L_0(0)$  through  $W L_{30}(0)$  have passed the erase verify. Word line  $W L_{31}(0)$  is coupled with erase pulse of -15V and -10V for the channel and edge erase operations, respectively. Word lines  $W L_0(0)$  through  $W L_{30}(0)$  are biased to the preferred voltages of 0V for channel erase, or +5V for edge erase, to reduce the disturb in the channel and edge erase operations respectively. The reduction in erase disturb is because the voltage drop across gate-drain and gate-source has been reduced from 5V to 0V for those cells such as  $M_a$ - $M_h$  shown in FIG. 23. The three concurrent word line voltages include 5V/0V that have been erase verified in Block 0, -10V/-15V on word line  $W L_{31}(0)$  and 0V/0V applied to non-selected word lines in non-selected blocks Block 1 through Block  $n$ , where the voltage designation is for edge erase/channel erase.

In FIG. 24 is shown a correction operation with preferred voltages for  $W L_m(n)$ , global  $B L_m$ , global  $S L(n)$ ,  $B T_1(n)$  and  $S T_1(n)$ . This operation is provided to correct those pages with over-erased cells and can be performed on bit-by-bit basis. For example, cells of  $M_b$  and  $M_c$  in word line  $W L_0(0)$  have been verified to be in an over-erase state requiring a  $V_t$  correction. The over-erase state is defined as the  $V_t$  below +0.5V. By contrast, cells of  $M_a$  and  $M_d$  have been erased successfully with  $V_t$  meeting the desired value which is to be above +0.5V but below 1.0V. The corresponding bit lines of  $B L_m$  and  $B L_{m+3}$  are coupled to +5V with  $B L_{m+1}$  and  $B L_{m+2}$  coupled to ground

to bias  $M_a$  and  $M_d$  in erase inhibit state with  $M_b$  and  $M_c$  in erase state. The rest of the word lines of  $WL1(0)$  through  $WL31(0)$  are applied with +2.5V to reduce the correction disturb in selected bit lines. Once  $WL0(0)$  is corrected successfully,  $WL0(0)$  will be switched to +2.5V and the operation is continued to the next word line  $WL1(0)$ .

This operation will be continued to the last page on word line  $WL31(0)$ . After the completion of the correction operation, all cells in Block 0 are "0" with  $V_t$  below +1V but above +0.5V. The three concurrent word line voltages required for this operation are  $V_{correction}$  for  $WL0(0)$  in Block 0, +2.5V for word lines  $WL1(0)$  through  $WL31(0)$  in Block 0, and 0V on non-selected word lines in non-selected blocks Block 1 through Block n.

FIG. 25 shows a correction verify operation with preferred voltages for  $WLM(n)$ , global  $BLM$ , global  $SL(n)$ ,  $BT1(n)$  and  $ST1(n)$ . This operation is provided to verify that pages with cells that have been corrected have a  $V_t$  below +1V and above +0.5V. For example, cells in  $WL0(0)$  can be verified collectively. The correction verify is same as a page read operation. Therefore, through transistors  $T0 - T3$ , all local bit lines  $B_n$  are connected to global bit lines  $BLM$  and sense amplifiers with a bias voltage around +1V. All source lines are held to ground by connecting local source lines  $S_n$  to  $SL(0)$  by means of transistors  $T4$  through  $T7$  gated by  $ST1(0)$ . As shown in FIG. 25, three preferred voltages are applied to all word lines concurrently. These three voltages include  $V_{corvfy}$  on the selected word line,  $WL0(0)$  for data verification, -4V on  $WL1(0)$ - $WL31(0)$  for shutting off bit line leakage and 0V on non-selected word lines in non-selected blocks Block 1 through Block n for avoiding any undesired disturbance.

FIG. 26 shows a random page program operation with preferred voltages for  $WLn(n)$ , global  $BLm$ , global  $SL(n)$ ,  $BT1(n)$  and  $ST1(n)$ . This operation is intended to change cell data of "0" to "1" on a page basis. The  $V_t$  of the cells is changed from +1.0V to more than +4V. In order to change cell data from "0" to "1" by using the cell bias conditions shown in FIG. 17a. In concert with  $V_{pgm}$  on word line  $WL0(0)$ , the corresponding bit lines of  $BLm$  and  $BLm+1$  are coupled to +5V to inhibit programming on  $Ma$  and  $Mb$  cells, and  $BLm+2$  and  $BLm+3$  are coupled to ground to enable programming on  $Mc$  and  $Md$  cells. The rest of the word lines of  $WL1(0)$  through  $WL31(0)$  are coupled with +2.5V to reduce the +5V disturb in selected bit lines. In this example, the data of  $Ma$  and  $Mb$  is kept to "0" but the data of  $Mc$  and  $Md$  is changed to "1." A preferred three voltages are required for this program operation. These three voltages include  $V_{pgm}$  on the selected word line,  $WL0(0)$  for channel program, +2.5V on  $WL1(0)$ - $WL31(0)$  for disturb reduction and 0V on non-selected word lines in non-selected blocks Block 1 through Block  $n$  for zero program disturb.  $V_{pgm}$  is approximately +15V in this operation.

Referring to FIG. 27, a random page program verify operation is shown with the preferred voltages for  $WLn(n)$ , global  $BLm$ , global  $SL(n)$ ,  $BT1(n)$  and  $ST1(n)$ . This operation is provided to verify those programmed cells in the selected page in the selected block. For example, cells in  $WL0(0)$  can be verified collectively. The random page program verify is same as page correction verify operation. The only difference is that  $V_{pgmvfy}$  is larger than  $V_{corvfy}$  used in the correction verify operation.  $V_{pgmvfy}$  is usually set to be around +4V for checking the  $V_t$  of programmed cells. This operation does not require three concurrent word line voltages.

The third embodiment of the present invention is explained with reference to FIG.28 through FIG.37. The cells are ETOX cells and are formed on a p-well. The p-well voltage is not always held to ground level as are the cells which are formed on a P-substrate. The detailed voltages for drain, gate, source and p-well nodes are shown in FIG. 28a through FIG.29d. The operations which use the three voltage word line of the present invention include block erase verify, erase inhibit, correction and correction verify. The operations of this embodiment are similar to the operations shown in FIG.8-FIG.15.

FIG. 28b shows a flash cell with bias conditions of the present invention illustrating an erase operation. The nodes D, G, S and Pw are coupled with +5V, -10V, +5V and +5V, respectively, for channel erase, where voltage values are exemplary values. The exact value and time of the control gate's voltage is subject to different flash technologies. The gate voltage of -10V, and source and drain voltage of +5V will result in a tunneling electric field in channel region of the cell. The tunneling electric field will transport electrons from the floating gate to the P-well in order to decrease the  $V_t$  of the cells (off-state) after a predetermined erase time. The erase operation can be performed on the basis of single-page (word line), block (N word lines), multiple blocks (M blocks) or chip (all blocks), where N and M are typically larger than two.

FIG. 28c shows a flash cell with bias conditions of the present invention illustrating an erase inhibit operation. The first erase inhibit is performed with D, G, S and Pw nodes set to +5v, +5V, +5V, +5V, respectively, and the second erase inhibit is performed with D, G, S and Pw nodes set to 5V, 5V/0V, 5V and 5V, respectively. This operation is intended to prevent a build-up of disturbance to those non-selected erased

cells (in either selected or non-selected blocks) and to achieve better endurance (number of program and erase cycles). Both gate voltages, 0V and 5V, are optional inhibit voltage. The "0V" creates less voltage stress on WL decoder but more disturbances. The "5V" has more voltage stress on WL decoder but less disturbances. The disturbance of both cases is too small to affect the value of  $V_t$  even during the specified maximum erase time.

FIG. 28d shows a flash cell with bias conditions of the present invention that illustrates the correction operation with D, G, S and Pw nodes coupled to +5V,  $V_{\text{correction}}$ , 0V, and 0V, respectively. The exact value and time for the control gate voltage in this operation varies with different flash technologies. The correction operation is a soft program CHE operation. The difference between program and correction lies in the control gate voltage. Normal program operation has about +10V applied to control gate and is intended to increase  $V_t$  to more than +5V. Correction has lower control gate voltage to avoid over program. It is used to correct the  $V_t$  of cells back to around +1V from either negative or below +0.5V to avoid sub-threshold leakage during subsequent read or program operations. The operation of FIG. 28d is sometimes referred to as recovery. The data becomes a logical "0" after this operation.

FIG. 28e shows a flash ETOX cell with the bias conditions of the present invention illustrating the CHE program operation with D, G, S and Pw nodes coupled to +5V,  $V_{\text{pgm}}$ , 0V and 0V, respectively. In the CHE program, there is a conduction current flowing from drain to source and causing an electron-hole pair generated at drain node. Electrons are attracted to the floating gate to increase  $V_t$  by the positive high voltage  $V_{\text{pgm}}$ , which increases the  $V_t$  of the cells. The CHE program typically consumes more

than 300uA per cell. With weak on chip charge pump circuits operating at Vdd below 3V, only about 4 bits can be programmed simultaneously. The cell data becomes a logical "1" after this operation is performed.

In FIG. 29a, a flash cell with bias conditions of the present invention illustrates the program verify operation with D, G, S and Pw nodes coupled to +1V, Vpgrmfy, 0V and 0V, respectively. The Vpgrmfy is an adjustable voltage input to the control gate of the cells to meet different Vt requirements in the program operation. For example, for storage of more than 2-bits per cell, Vpgrmfy may vary from as low as 3V up to about 5V or more. For a binary program, Vpgrmfy is set to be around +5V. For storage of multiple states such as 1V, 2V, 3V, and 4V, Vpgrmfy is set to 1V, 2V, 3V and 4V accordingly to verify each state

In FIG. 29b, a flash cell with bias conditions of the present invention illustrates the correction verify operation with D, G, S and Pw nodes coupled to +1V, Vcorvfy, 0V and 0V, respectively. The Vcorvfy is an adjustable voltage input to the control gate to meet different Vt requirements in this operation. The Vt of over erased cells will be recovered back to a Vt-window of between + 0.5V and 1.0V, after this operation is performed.

FIG. 29c shows a flash cell with bias conditions of the present invention illustrating a read operation with D, G, S and Pw nodes coupled to +1V, Vread, 0V and 0V, respectively. The Vread is an adjustable voltage input for the control gate for the read operation. Vread can be simply set to Vdd; however, in some designs, Vread is set to a clamped value so that the read voltage applied to the control gate can be

independent of  $V_{dd}$  variation. In some designs, the  $V_{read}$  voltage is boosted to be higher than  $V_{dd}$ , resulting in higher read current.

Shown in FIG. 29d is a flash cell with bias conditions of the present invention illustrating an erase verify operation with D, G, S and Pw nodes coupled to +1V,  $V_{ersvfy}$ , 0V and 0V, respectively. The  $V_{ersvfy}$  is an adjustable voltage input to the control gate for this operation. In the conventional ETOX cell,  $V_{ersvfy}$  is set to be around +2.5V to reduce the number of over erased cells because the correction cannot be performed in truly bit-by-bit mode as in regular program operation. The correction is done in a collective mode. When the number of over-erased cells increases to some level, the correction current will overload the on-chip charge pump and fail to recover the  $V_t$  of the cells. In contrast,  $V_{ersvfy}$  is set to be +1V in the present invention and as a result of the 3-level word line voltage is used to perform bit-by-bit correction. There are many cells in many bit lines, but only one cell per one selected bit line is corrected simultaneously. Therefore, current over load will not occur and the corrected  $V_t$  can be set accurately.

In FIG.30 is shown a conventional ETOX NOR type flash EEPROM memory array formed on a P-well 30. This nonvolatile NOR-type memory array includes: a matrix of word lines and bit lines intersecting one another; and an ETOX memory cell being disposed so as to correspond to each intersection of the matrix of the global bit lines  $BL_n - BL_{n+1}$ , local bit line  $B_n$ , source line  $SL_n$  and global word lines  $WL_n$ . The memory cell including a control gate; a drain, a source and a P-well is as shown in FIG. 28. The control gates are coupled to a corresponding one of the row wise word lines  $WL_m(n)$ , the drains are coupled to a corresponding one of the local column wise bit line

B<sub>n</sub> and one of the global bit line BL<sub>n</sub> selected by transistor T<sub>n</sub> gated by BT1 (n) and BT2(n), and the sources are coupled to a corresponding one of the local row wise source lines SL<sub>m</sub>. The memory cell is capable of performing a FN erase and a CHE program operation based on the 3-level word line of the present invention. The plurality of control signals of WL<sub>m</sub>(n), BT1 (n) and BT2(n) are generated from an X-decoder (word line decoder), local bit line decoder, local source line decoder, global bit line decoder and global source line decoder, respectively.

Shown in FIG. 31 is a block erase operations for a conventional ETOX NOR type flash EEPROM memory array formed in a P-well 41 and 42. The block erase is an channel erase operation. For a channel erase operation, an erase voltage of -10V is coupled to the corresponding row wise word lines, WL0(0)-WL511(0) for selected Block 0, and ground is coupled to the non-selected word lines in the rest of the blocks. The drains of the cells are left floating by the corresponding local column wise first level metal bit lines B0-B3 and the second global metal bit lines BL<sub>n</sub>-BL<sub>n</sub>+1 via transistors of T0 and T1 which are gated by applying 0V to BT1 (n) and BT2(n) signals. The sources are coupled to 5V by a corresponding plurality of row wise source lines SL(0). The P-well voltage for Block 0 is set to +5V, and the P-well for the blocks not being erased, Block n is set to 0V. The channel erase conditions are applied so that the memory cells of Ma -Ml in the word lines in Block 0 are capable of performing a FN erase operation. The V<sub>t</sub> of the cells are decreased after the erase operation is performed. The flash cells of Mm-Mx in non-selected Block n are kept undisturbed. The erase operation is designed to be an iterative process. Each erase pulse width can be set to around 1 ms.



Given a shorter erase pulse, a larger number of erase pulses are required. Three concurrent word line voltages are not required in this block erase operation.

In FIG. 32 is illustrated a block erase verify operation for an ETOX NOR array formed in a P-well with preferred voltages for  $WLn(n)$ , global  $BLn$ , global  $SL(n)$ ,  $BT1(n)$ , and  $BT2(n)$ . Assuming  $WL0(0)$  is first selected for Block erase, then  $Versvfy$  is coupled to  $WL0(0)$ . The rest of the word lines of  $WL1(0)$  to  $WL511(0)$  are coupled to  $-4V$  to shut off any potential leakage caused by over-erased cells that might exist in Block 0. The word line voltage of  $-4v$  is not a fixed number but is set to be able to shut off any leakage current resulted from cells with a  $V_t$  less than  $-4V$ . All source voltages  $SL(n)$  are coupled to ground. Bit line  $BLn$  is coupled to  $0V$  and  $BLn+1$  is coupled to  $1V$  for the verify operation. Bit line  $BLn+1$  is then selectively connected to a corresponding sense amplifier for verification. The P-wells 41 and 42 for all blocks are coupled to  $0V$ . The process is continued in the same page until all cells in the page are verified. Then the verify process is moved to the next page of  $WL1(0)$  in Block 0. After five hundred and twelve word lines are verified, the block erase verify is terminated. With the successful verification of block erase all cells in the selected block (Block 0) for erase operation become a logical "0". The data for cells in the non-selected blocks remain the same without changes. Although the cells in Block 0 are at a logical "0", there could be over erased cells. The definition of an over erased cell in the present invention is a cell  $V_t$  ranging from a negative  $V_t$  to a positive  $V_t$  but below  $0.5V$ . The over-erase cells will induce leakage and result in false readings so that a  $V_t$  correction operation is required. The three concurrent word line voltages,  $Versvfy$ ,  $-4V$  and  $0V$  are required for this operation.

Fig. 33 shows an erase inhibit operation for an ETOX NOR array formed on a P-well with preferred voltages for  $WLM(n)$ , global  $BLn$ , global  $SL(n)$ ,  $BT1(n)$ , and  $BT2(n)$ . This operation is performed on sub-block basis and is intended to set those sub-blocks that have been successfully erased into a de-selected mode to prevent further erase. For example,  $WL0(0)$  and  $WL1(0)$  are verified to have a successful erase and are set to be in Erase Inhibit mode to avoid the further erase pulses. The way erase inhibit is set, the word line voltage is set to +5V/0V for channel erase with the drain floating and the source voltage at +5V. In the erase inhibit mode the tunneling electric field is reduced so that no tunneling effect will take place. The P-well voltage for the active block, Block 0, is +5V, and the P-well voltage for the other blocks not having an erase operation is 0V. Three concurrent word line voltages are not required for this operation; however, if the operation is used in conjunction with an erase operation there would be three concurrent word line voltages, -10V for word lines being erased, +5v for inhibiting word lines in the selected block and 0v for word lines in non-selected blocks Both gate voltages, 0V and 5V, are optional inhibit voltages. The "0V" causes less voltage stress on WL decoder but more disturbance. The "5V" causes more voltage stress on WL decoder but less disturbance. The disturbance of both cases is too small to affect the value of  $V_t$  even during spec maximum erase time.

In FIG. 34 a correction operation is illustrated with preferred voltages for  $WLM(n)$ , global  $BLn$ , global  $SL(n)$ ,  $BT1(n)$  and  $BT2(n)$ . The correction operation is performed on a bit-by-bit basis and is intended to correct those over erased cells to a  $V_t$  voltage that is above +0.5V but below +1.0V. The process is repeated to correct all cells in one selected word line  $WL0(0)$  and then moved to correct the cells in next word lines of

Block 0. A Vcorrection voltage is coupled to the first selected word line WL0(0) along with a bit line voltage of 5V in order to perform a CHE soft program. The Vcorrection voltage is set to be less than +10V while the rest of the word lines of WL1(0) to WL511(0) are coupled to -4V to shut off any potential leakage due to over erased cells that might be existing in Block 0. The -4V is an approximate value and is of sufficient magnitude to be able to shut off any leakage current resulting from cells with  $V_t$  less than -4V. All source voltages of SL(0)-SL(n) are coupled to ground along with bit line BLn. Bit line BLn+1 is couple to 5V for the correction operation. The P-well of all blocks is couple to 0V. The process is continued in a same page until all cells in same page are corrected and then the process moves to correct next page WL0(1) in Block 0. After all word lines in a block are successfully corrected, the correction process is terminated. With the successful correction, all cells in the selected block (Block 0) become "0". The data of cells in the non-selected blocks remain the same without changes. This operation requires the use of three concurrent word line voltages, Vcorrection, -4V, and 0V.

FIG. 35 illustrates a correction verify operation with preferred voltages for WLn(n), global BLn, global, SL(n), BT1 (n), BT2(n). This operation is performed on bit-by-bit basis. In the present invention, this operation is intended to verify that those over erased cells are corrected to  $V_t$  within +0.5V but below +1V. A voltage Vcorvfy is coupled to word line WL0(0) and the rest of the word lines, WL1(0) to WL511(0), are coupled to -4V to shut off any potential leakage caused by over erased cells that might be existing in Block 0. The voltage of -4v is an approximate value and is set to be able to shut off any leakage current which results from cells with  $V_t$  less than -4V. All source

voltages of  $SL(0)$ - $SL(n)$  are coupled to ground along with the bit line  $BLn$ . The bit line  $BLn+1$  is coupled to +1V for the verify operation. The P-well of all blocks is coupled to 0V. The process is continued in the same page until all cells in that page are corrected and verified. Then the operation moves to verify next page of  $WLn(0)$  in Block 0. After all word lines in the block are successfully corrected and verified, the verify process is terminated. This operation requires three concurrent word line voltages,  $V_{corvfy}$ , -4V and 0V.

In FIG. 36 is illustrated a CHE program operation with preferred voltages for  $WLn(n)$ , global  $BLn$ , global  $SL(n)$ ,  $BT1(n)$ , and  $BT2(n)$ . This operation is performed after correction and is on a bit-by-bit basis. The process is continued until all cells are fully verified. A voltage  $V_{pgm}$ , which is approximately 10V, is coupled to  $WL0(0)$ , and the remainder of the word lines of  $WL1(0)$  to  $WL511(0)$  are coupled to 0V. Because all over erased cells have previously been corrected, the -4v used to shut off any leakage is no longer needed. The CHE program operation continues on a bit-by-bit basis and terminates when all cells in same byte/word that are programmed, coupled to a high  $V_t$  (>4V).. This CHE program operation does not require three concurrent word line voltages.

In FIG. 37 a CHE program verify operation is illustrated with preferred voltages for  $WLn(n)$ , global  $BLn$ , global  $SL(n)$ ,  $BT1(n)$ , and  $BT2(n)$ . The P-well voltage is set to 0V. This operation is performed in a similar manner as program verify in FIG. 15. This program verify operation does not require three concurrent word line voltages.

The fourth embodiment of the present invention will be described with reference to flow charts of FIG.38 to FIG.44 to illustrate the three concurrent

word line voltages technique of this invention. In FIG. 38a is shown the  $V_t$  distribution of ETOX cells of prior art for a large block after performing FN block erase and CHE block program operations. Prior to the erase operation, the data of the cells contained both "1" and "0" data of different  $V_t$  voltages. All cells are pre-programmed to high  $V_t$  state above  $V_{t1}$  as shown in waveform 600. Typically,  $V_{t1}$  is approximately +4V for a binary-data cell. Since the pre-program operation is performed on a bit-by-bit basis, the  $V_t$  distribution of the "1" for programmed cells can be controlled to be very narrow. The flash cells with  $V_t$  larger than  $V_{t1}$  store "1" data.

Continuing to refer to FIG. 38a, subsequent to the pre-programming, an FN erase operation is performed on a block basis to lower  $V_t$  of all programmed cells. This operation is to change the cell data from "1" to "0" in the selected block and creates a distribution as seen in waveform 500. After block erase, the  $V_t$  of the cells in the selected block are brought lower and below  $V_{t0}$ . The cells with a  $V_t$  below  $V_{t-1}$  are referred to as over-erased cells 501 that require a correction 502. In the NOR type flash array, the voltage  $V_{t-1}$  is the lowest acceptable positive  $V_t$  (not negative) to guarantee there is no occurrence of the over erase problem 501 for normal operation. Since the erase operation is performed on block basis and is less controllable than the program operation done on a bit-by-bit basis. As a result, the  $V_t$  distribution of "0" data 500 after block erase operation is much broader than the "1" data. The voltage  $V_{t-1}$  in FIG. 38a is approximately +0.6V,  $V_{t0}$  is approximately +2.5V and  $V_{t1}$  is approximately +4V for 3V operation.

Referring to FIG. 38b, a  $V_t$  distribution is shown of ETOX cells in a large block when the three concurrent word line voltages of the present invention are used. As seen

from waveforms 550 and 650, the  $V_t$  distributions of the programmed state of data "1" and the block erased state of data "0" can be made very narrow. The detailed explanation to show how to achieve the narrow-distribution of data "0" and data "1" is with reference to the flow diagram of FIG. 39. Over erased cells 551 are corrected 552 using a correction operation discussed with reference to the flow diagram on FIG. 40.

FIG. 39 shows a simplified flow diagram of a block erase operation when the present invention is applied to an ETOX cell. The block erase operation starts with selecting one block 51 which sets up a plurality of addresses for an X-decoder and Y-decoder to select the right block. This is then followed by a FN block erase operation 52 to decrease the  $V_t$  of the selected cells in the selected block. The erase operation is an iterative process, and each time an erase pulse of a pre-determined time is performed on the selected block, all erased cells are verified to determine if the  $V_t$  value is below  $V_{t0}$  53 on a sub-block basis. Any sub-blocks that have been verified to satisfy that  $V_t$  is below  $V_{t0}$  are immediately set to an inhibit state by sub-block erase inhibit 54 to avoid further erase operation. The process continues to check determine if all sub-blocks are verified 53. The process stops when all pages of the sub-block are verified successfully 55. Any sub-block that is not inhibited 56 is returned to the erase operation 52. In the process shown in the flow diagram of FIG 39, three concurrent applied word line voltages are used in some steps. The first is in decision step 53 (Versvfy, see FIG. 10 and FIG. 32) and the second one is in step 54 see FIG. 33).

FIG. 40 shows a preferred flow diagram for a correction operation of the present invention. This operation is intended to correct those over erase cells in an erased block. Any cells with  $V_t$  below  $V_{t-1}$  must be corrected to a voltage above  $V_{t-1}$ . With

three concurrent word line voltages, this operation can be performed on a bit-by-bit basis. It should be noted that correction is performed byte by byte, but correction will depend on the threshold voltage of each individual bit in the byte. The purpose is to avoid any over-correction. Thus, "bit-by-bit" means that the correction inhibition will stop correction on any successfully corrected bit. Therefore, a narrow  $V_t$  distribution of data "0" is achieved. The first step is to perform a check 57 is to verify that  $V_t < V_{t-1}$ . If the  $V_t$  of any cells is below  $V_{t-1}$ , the cells are corrected 58 to increase the  $V_t$  of those cells. Since step 57 is carried out on bit-by-bit basis, all bytes in the selected block are checked 59. The next byte 60 is verified until all pages in selected block have been verified to have a  $V_t > V_{t-1}$ . Whenever an over erase byte is found, an immediate byte correction is performed 58 and  $V_t$  is increased above  $V_{t-1}$  for the over erased byte. The reason for byte-correction is that an ETOX cell which uses CHE correction which consumes too much power and can not be performed on page basis. Each page contains a plurality of bytes, and the byte correction operation is performed consecutively byte by byte until all bytes in the selected page are corrected. The correction operation uses three concurrent applied word line voltages,  $V_{\text{correction}}$  for byte correction 58 and  $V_{\text{corvfy}}$  for byte correction verification 57.

FIG. 41a shows the  $V_t$  distribution for AND cells in prior art of a large block after a FN block erase and a FN page program operations on all pages in a block are performed. Prior to the erase operation, the data of the cells contained both "1" and "0" at different  $V_t$ s. All cells were erased to high  $V_t$  state (above  $V_{t1}$ ) as shown in waveform 300. Typically,  $V_{t1}$  is around +4V for a binary-data cell. Since the erase operation is performed on block basis, the  $V_t$  distribution of "1" data 300 of the erased cells can not

be controlled well. As a result, a wide  $V_t$  distribution of "1" data 300 is generated in prior art. Subsequent to the erase operation, a FN page program operation is performed to lower  $V_t$  of all erased cells. This operation changes the cell data from "1" to "0" in the selected block as seen in waveform 200. After the page program, the  $V_t$  of the cells is brought lower below  $V_{t0}$ . Cells with a  $V_t$  below  $V_{t-1}$  are defined as over-erased cells 301 that require  $V_t$  correction. In a NOR type flash array, the voltage  $V_{t-1}$  is the lowest acceptable positive  $V_t$  (not negative) in order to guarantee no occurrence of the over erase problem for a normal operation. Since the program operation is performed on page basis, it is more controllable than the erase operation which is performed on a block basis. As a result, the  $V_t$  distribution of "0" data 200 is much narrower than the distribution of "1" data 300. The voltage  $V_{t-1}$  is approximately +0.6V,  $V_{t0}$  approximately +1.0V and  $V_{t1}$  is approximately +4V for a 3V operation.

In FIG. 41b is shown a  $V_t$  distribution for AND cells of second prior art. Unlike the approach in FIG. 41a, the program operation is to increase the  $V_t$  of the cells 200 and is carried out on page basis for all pages in a block. The erase operation decreases the  $V_t$  of the cells 300 and is carried out on block basis. Therefore, "1" data has narrower  $V_t$  distribution than "0" data because there is more control over the operation carried out on a page basis. The  $V_{t0}$  in FIG. 41b is set to be much higher than  $V_{t0}$  in FIG. 41a, because the cell program is performed on page with respect to the prior art of FIG. 41a, and the erase operation in FIG. 41b is performed on block basis. The prior art of FIG. 41a uses a bit-by-bit program scheme to obtain "0" data so that over-erase will not occur. The  $V_{t0}$  can be set around +1.0V with  $V_{t-1}$  of +0.5V in FIG. 41a. The prior art of FIG. 41b uses block erase, and the erase operation cannot be performed on a bit-by-



bit basis. In FIG.41b,  $V_{t0}$  is set at higher value +2.5V to reduce the number of over-erased cells.

FIG. 41c shows the  $V_t$  distribution for cells of the present invention and having a tight distribution for "1" data 250 and "0" data 350. All cells in a selected block are first collectively erased below  $V_{t0}$  by FN tunneling as shown in distribution 350. The voltage  $V_{t0}$  is set to be approximately +1.0V for the present invention. Any over erased cells 351 with a  $V_t$  below  $V_{t-1}$  (+0.5V) is corrected 352 back above  $V_{t-1}$  but below  $V_{t0}$  by means of the bit-by-bit correction operation of this invention.  $V_{t0}$  is used for block erase verify and  $V_{t-1}$  used for page correction verify. Although erase is performed on block basis, the post-erase, bit-by-bit correction makes a very tight  $V_t$  distribution for "0" data 350 in the present invention. A FN channel program operation is performed to increase  $V_t$  of selected cells on a page basis for all pages in a block. This operation changes the data of the cells from "0" to "1" in the selected block and results in the distribution of threshold voltages as seen in waveform 250. After the FN channel program, the  $V_t$  of the cells are raised above  $V_{t1}$ .

Continuing to refer to FIG. 41c, in a NOR-type flash array, the  $V_{t-1}$  voltage is the lowest acceptable positive  $V_t$  to guarantee no occurrence of the over erase problem in a normal operation. Since the program operation is performed on page basis for all pages in a block, it is more controllable than the block erase operation. As a result, a tight  $V_t$  distribution of "1" data 250 is achieved. By using the techniques of the present invention, both tight  $V_t$  distribution of "1" data 250 and "0" data 350 can be attained.

FIG.42a shows a simplified flow diagram of a write operation for an AND cell.

The write operation includes a block erase 20 to increase the  $V_t$  of cells above  $V_{t1}$

collectively and a page program 21 to selectively decrease the  $V_t$  below  $V_{t0}$ . FIG.42b shows a simplified flow diagram of a write operation for a second AND cell. The write operation includes a block erase 22 to decrease  $V_t$  of the cells collectively below  $V_{t0}$ , and programming 23 selectively increase  $V_t$  above  $V_{t1}$ . In either FIG.42a or FIG.42b, block erase and program are based on concurrent two voltage word line designs,  $V_{t0}$  and  $V_{t1}$ .

FIG.43 shows a simplified flow diagram of a block erase operation for an AND-like cell of the present invention. This flow diagram starts by selecting a block 61 for erase. Then a block erase 62 is performed where the  $V_t$  of the cells in the block is decreased. After the first block erase pulse is executed on all selected pages, a page verification 63 is performed to check if  $V_t$  is below  $V_{t0}$ . If the selected page meets the criteria  $V_t < V_{t0}$ , it will be set to the erase inhibit state 64 to prevent further erasing; otherwise, the verification is continued to set any next pages in erase inhibit. The page verification will stop when all pages 65 are verified and set to page erase inhibit. Subsequently, a second erase pulse is selectively applied to those pages not in the state of page erase inhibit. The operation is branched back to block erase 62 where  $V_t$  is decreased when all pages are not in erase inhibit. Each time, the number of erased word lines will be reduced in the block erase operation 62. The operation will terminate when all pages are set into page erase inhibit. Unlike the method in FIG.42a and FIG.42b, the three concurrent applied word line voltage technique of the present invention is used in the block erase operation shown in FIG. 43, verification of one page cells 63 (Versvfy), and page edge erase inhibit 64 (+5V/0V, see FIG.23).

FIG. 44 shows a preferred flow diagram for a correction operation of the present